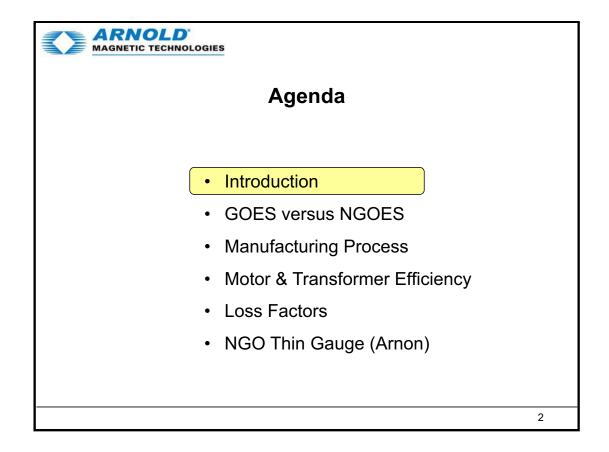
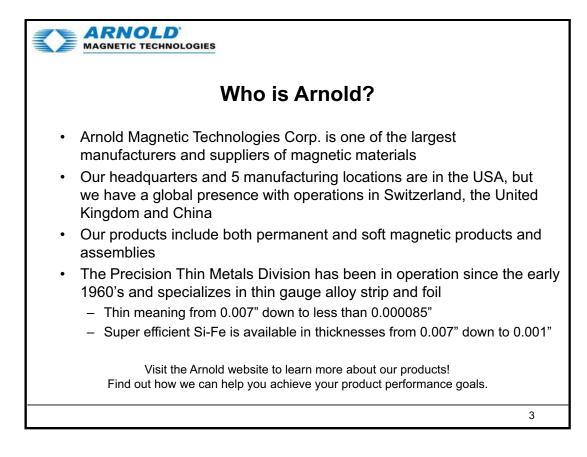


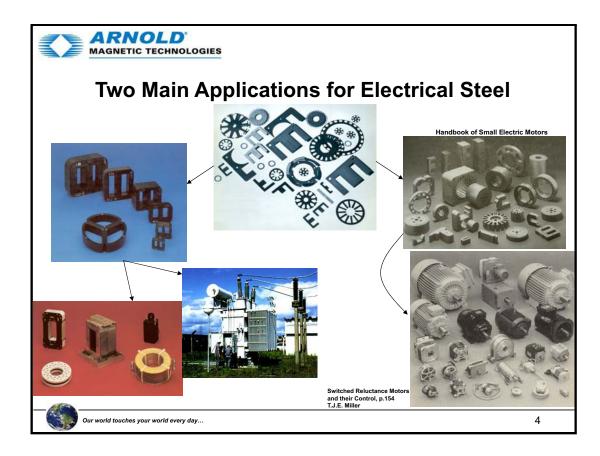
- Demand for electrical device efficiency gains which started decades ago due to economic factors have accelerated, propelled by the unprecedented rise in energy costs and the alleged impact of energy consumption on global weather instability.
- Arnold's thin gage silicon-iron products, both grain-oriented and non-grain oriented (Arnon) offer substantial advantages in many applications.



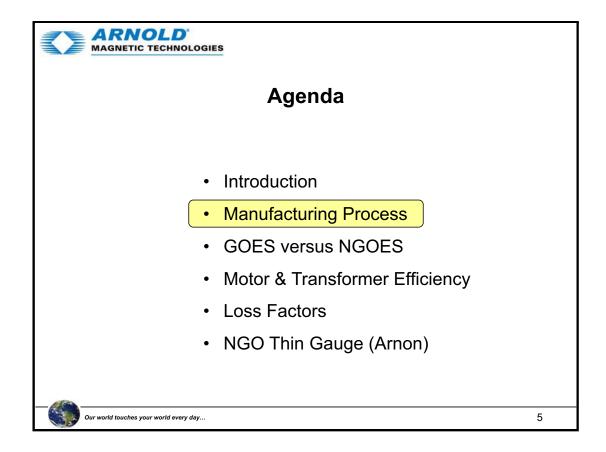
- Over the last several years I've spoken about permanent magnets. However, the magnetic material that is utilized in far great tonnage than magnets is the family of soft magnetic materials, especially Silicon-Iron whose usage spans both motor/generator and transformer/inductor applications.
- Often referred to as Electrical Steel, Silicon-Iron (Si-Fe) is available in both Grain Oriented (GOES) and Non-Oriented (NGOES) forms.



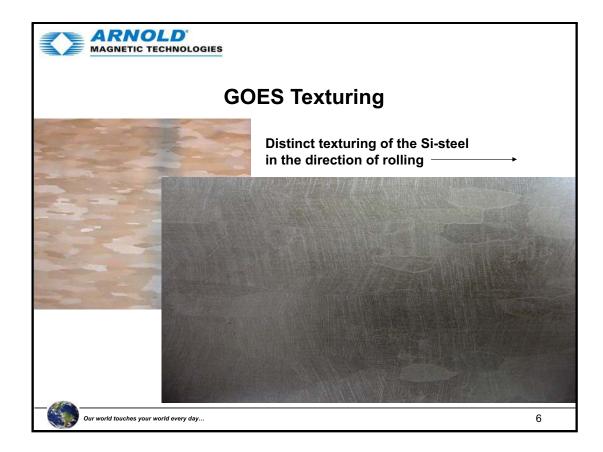
- Arnold started as a privately owned manufacturer over 100 years ago in northern Illinois.
- The founder, Bion Arnold, was a world famous electrical engineer and inventor. His son, Robert, took over the running of the company until the late 1960's.
- In the 1950's, Arnold was sold to Allegheny Steel Company (Allegheny Ludlum, Allegheny International) and remained part of Allegheny until 1986.
- It was during the Allegheny ownership period that Arnold developed its rolling mill capabilities and many of the electrical steel grades such as Arnon 5 and Arnon 7.
- Product development continued and today Arnold's Rolled Products Division provides the highest available quality strip and foil in dozens of materials, both magnetic and non-magnetic.



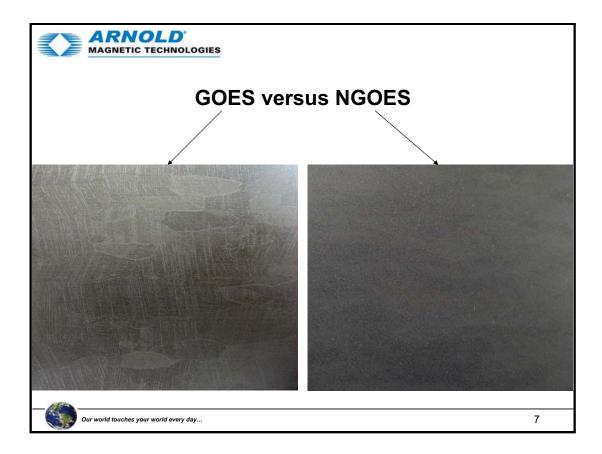
- There are two major uses for Electrical Steel: Transformers/Inductors and Motors/Generators.
- Transformers are used not only for Electrical Power Distribution, but in virtually every power supply in every electrically powered electronic device in offices and homes.
- Motors have moved from industrial uses in the 1800 and early 1900's to home and automotive uses. In homes for example, we use motors in garage door openers, washing machines and dryers, furnaces for home heating, refrigerator compressor pumps, bathroom exhaust fans, garbage disposals, and electric powered shop tools.
- Motor usage in automobiles has expanded to include windshield wiper motors, seat adjusting motors, starters, cooling fans, passenger compartment fans, alternators, etc. Note however, many of these use low carbon steel cans for low cost with laminations reserved for higher output, higher efficiency motors/generators.
- The newest example of lamination steel use in cars is the electric drive motor of hybrids or full electric vehicles.



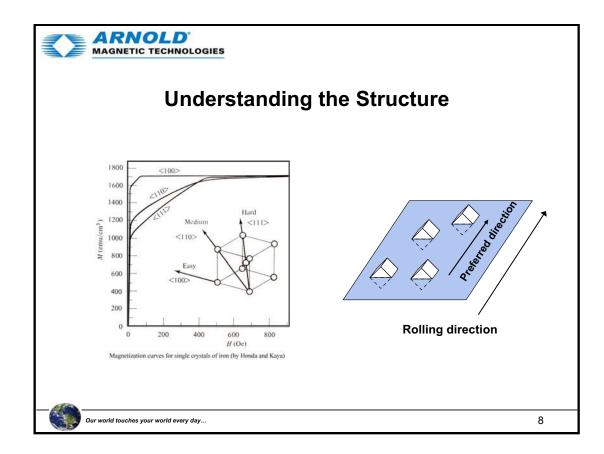
- We'll briefly discuss the manufacturing processes for Grain Oriented and Non-Grain Oriented Electrical Steel.
- Arnold's name for NGOES is Arnon.
- In order to achieve the especially advantageous properties of Arno n requires at least 10 manufacturing steps.
- To understand why these are necessary requires a basic understanding of the materials.



- Silicon-Iron consists of BCC (body-centered cubic) crystals which, during rolling, are stretched and flattened. If they are left in that state, the magnetic properties are maximized but only in the rolling direction, that is along the length of the strip.
- In these two photos we can see the elongated and flattened structure of Grain Oriented Electrical Steel (GOES).



- However, when a final annealing heat treat is performed, nearly isotropic properties are achieved within the plane of the strip.
- This is exemplified by the almost total elimination of the grain structure as seen in the photo to the right.



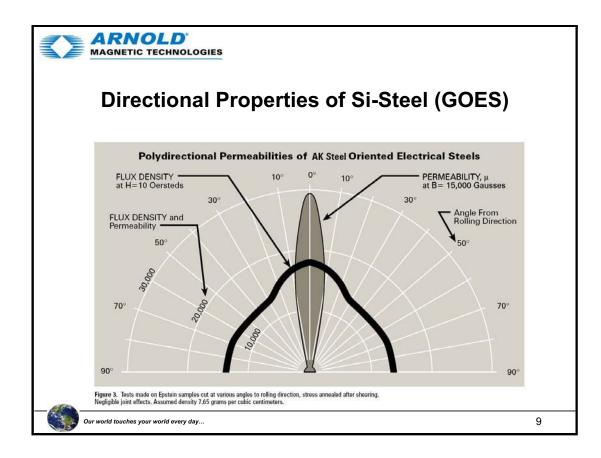
- By the 1940's it was well known that the iron single crystal has a preferred direction along which magnetization is easier than the others the Goss Structure**. This translates to higher permeability and lower core loss in that given direction, the "cube-on-edge" direction.
- It was recognized that if all the grains in a lamination can be aligned this way, the lamination can be cut to take advantage of this performance enhancement.
- This structure and process were developed to create the preferred orientation in the rolled silicon steel strip. The process involves a combination of cold working and heat treating to produce a strip where most of the metal grains are oriented along the rolling direction.
- While the material is not uniform in all directions, many components such as wound cores and stamped/stacked E cores can be made to maximize the positive aspects of orientation.
- As shown in the illustration at the left, the eas axis (of magnetization) is along the <100> crystal axis.
- The "hard" axis is from one corner to the diagonally opposite corner: the <111> crystal axis.
- To reiterate, effective application of the material requires using the strip in the same direction that flux is expected.

He graduated from Case Institute of Technology in 1925. He made significant contributions to the field of metals research and in 1935 he patented a method[1] and published a paper[2] describing a method of obtaining so-called grain-oriented electrical steel, which has highly anisotropic magnetic properties. This special "grain-oriented" structure was named after its inventor and it is referred to as the "GOSS structure".

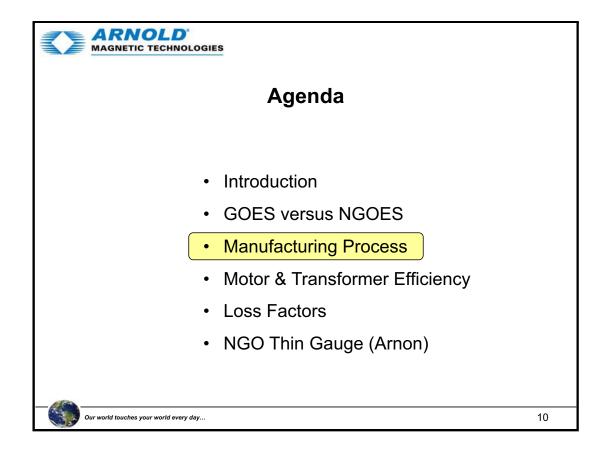
The grain-oriented electrical steel allowed development of highly efficient electrical machines, especially transformers. Today, the magnetic cores of all high-voltage high-power transformers are made from grain-oriented electrical steel.

-- Wikipedia --

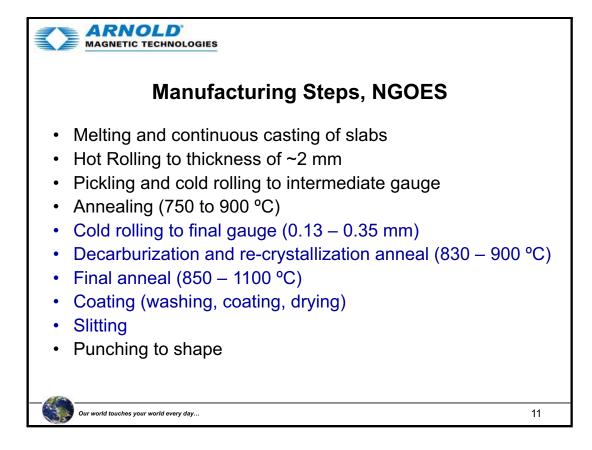
^{**}Norman P. Goss (born February 04, 1906 - died October 28, 1977) - inventor and researcher from Cleveland, USA.



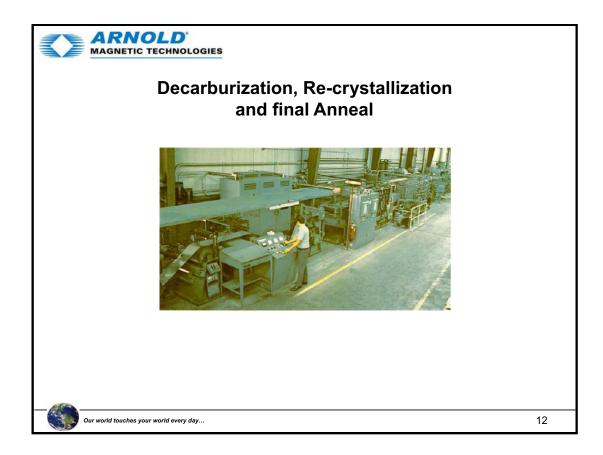
- This chart from AK steel for grain oriented silicon iron illustrates the difference in magnetic properties in the rolling direction (straight up and down) and at angles at up to 90° from the rolling direction (left to right).
- Maximum performance is achieved by application of the material with the strip in the same direction that flux is expected.
- In contrast, NGOES shows only a 10-20% variation in properties within the plane of the strip.



- Reduced eddy current related loss in thin gauge lamination materials such as Arnold's grades (between 0.001 and 0.007" thick) permit higher frequency operation.
- Silicon-Iron laminations are more expensive than low carbon steel and their use has been reserved for high energy density, high rpm motors and generators and for high frequency transformers.
- In defense of Silicon-Iron manufacturers, the process for making lamination-ready Si-Fe strip involves many processing steps and utilizes expensive manufacturing equipment.



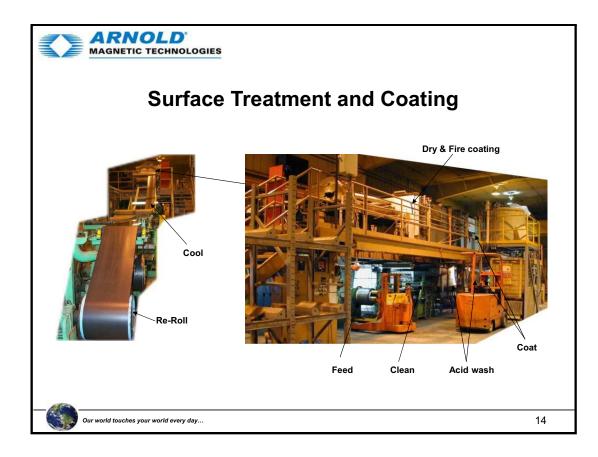
- There are at minimum 10 manufacturing stepsutilizing equipment such as melting furnaces, hot and cold rolling mills, atmosphere treatment furnaces, acids and inorganic coating materials, tool steel precision slitters, etc.
- Some photographs of the manufacturing steps highlighted in blue are on the following slides.



- Re-crystallization and annealing is performed in continuous atmosphere furnaces such as this one at Arnold.
- The strip can be seen exiting the furnace at the left of this photo...



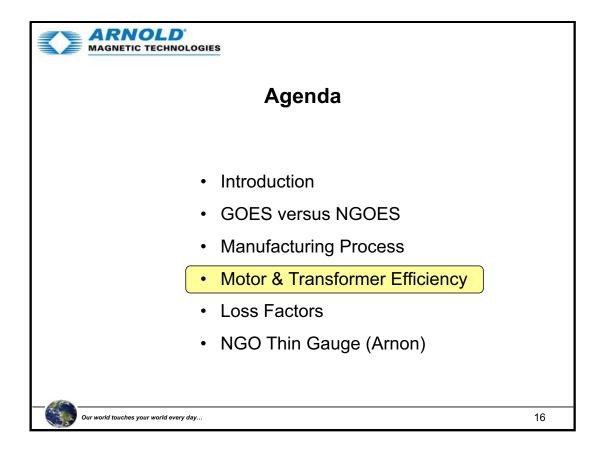
- For thin gauge, specialized rolling mills are used to achieve uniform cross-strip thickness and flawless surfaces on Si-Fe down to 0.001" (0.025 mm).
- Arnold does, incidentally, roll other alloys to as little as 85 millionths of an inch (0.000085") on Sendzimir mills.
- The Sendzimir mill uses small diameter rolls to reduce the strip thickness. The small rolls are backed-up by larger rolls to eliminate bending of the small rolls under intense compressive stress.
- The use of small diameter inner rolls is required to apply the very high contact pressures required to compress the sheet thickness while it is being stretched by the powerful drive and take-up motors.



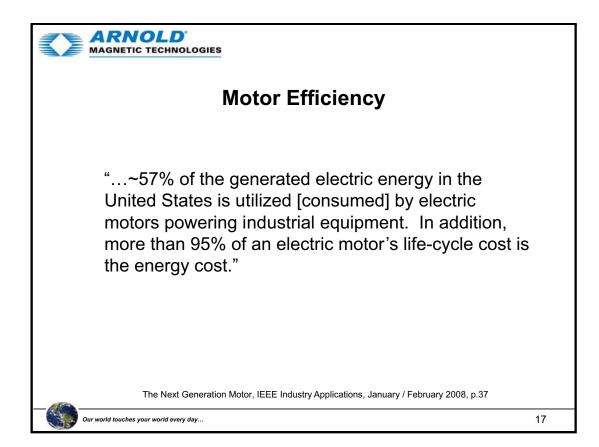
- Surface coating includes steps of acid wash, coating and firing the coating onto the strip.
- The coating system shown here is two stories high and approximately 120 feet long.
- Typical coating for Si-Fe laminations is designated "C-5".

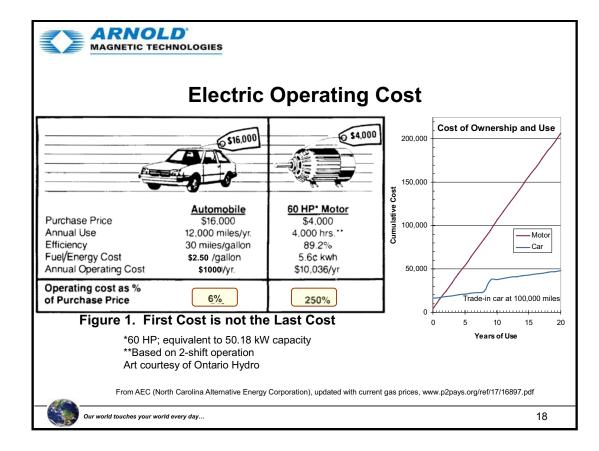


- While it may be more efficient to process wide strip, applications often require a narrow section either to match the device size or to maximize utilization.
- Significant material loss occurs when the center of a round lamination is punched-out.
- Many larger devices benefit from segmenting the laminations and using an alternated stacking arrangement to: 1) maximize material usage and 2) minimize reluctance losses due to gaps in the lamination stack.



- Most commercial transformers use low-cost low carbon steel.
- Justification for the use of Si-Fe includes demands for higher device efficiency (motors and transformers) and design feasibility (very high frequency transformers).

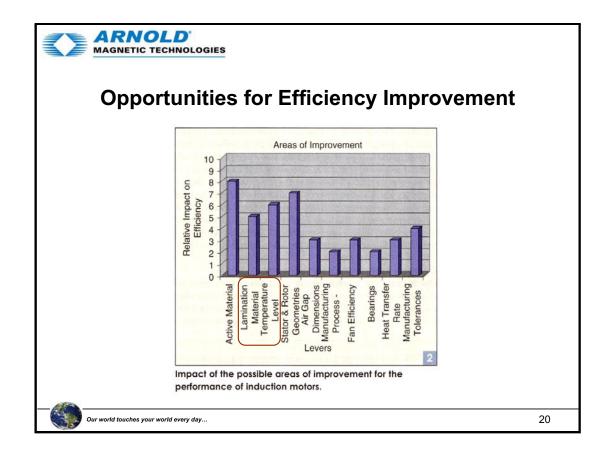




- This chart from AEC has been updated with more current gasoline pricing. Electric costs are likely to rise commensurate with gas but have not been changed here.
- It is obvious that more attention must be paid to total operating costs, not just initial purchase price. Our industry needs to help educate the customer to consider more than just initial purchase price.

Mot	tor Efficiency
Perpetual Motion	Efficiency on 100%
Super Conducti	ting Motor 98.5–99.5%
Amorphous Ste	eel Laminations 98–98.5%
Potential Coppe	ver Rotor 96.5–97.5%
Today's Premiu	um Efficiency 96.2%
EPACT	95%
1975	<94%
200, HP Motor - Pa	ast, Present, and Future
Typical officiency los	evels over time.

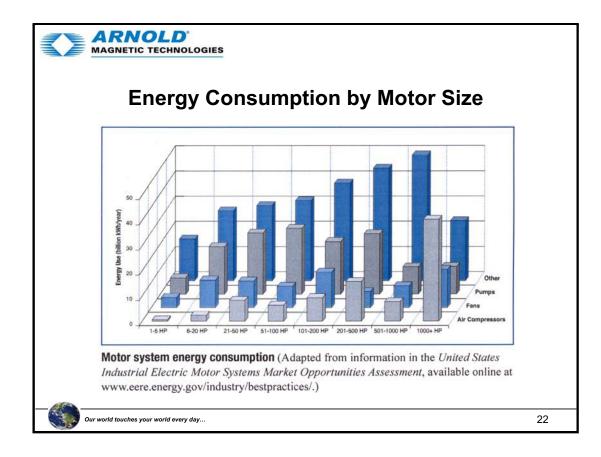
- In 1975 the average efficiency of large electric motors was less than 94% with many operating at 80% or less.
- Today's premium motors achieve over 96% efficiency and the government is mandating further increases.



- Efficiency increase is affected by many facets of the motor. One estimate is that close to 5% of total energy loss is due to lamination materials.
- With targets of 96% plus efficiency, it is imperative to optimize lamination material.
- · Lossy material adds to heat buildup which also reduces motor life expectancy.

	Loss	Distrik	oution
СОМРАН			PERCENT FOR MOTORS TESTED
and the second	IN TH	E EASA/AEMT S	TUDY [12].
Losses Core losses (W _c)	Average 19%	Average 21%	Design Factors Affecting Losses Electrical steel, air gap, saturation, supply frequency, condition of interlaminar insulation
Friction and windage losses (W _{fw})	25%	10%	Fan efficiency, lubrication, bearings, seals
Stator I2R losses (W _s)	26%	34%	Conductor area, mean length of turn, her dissipation
Rotor I2R losses (Wr)	19%	21%	Bar and end ring area and material
Stray load losses (Wi)	11%	14%	Manufacturing processes, slot design, air gap, condition of air gap surfaces and end laminations

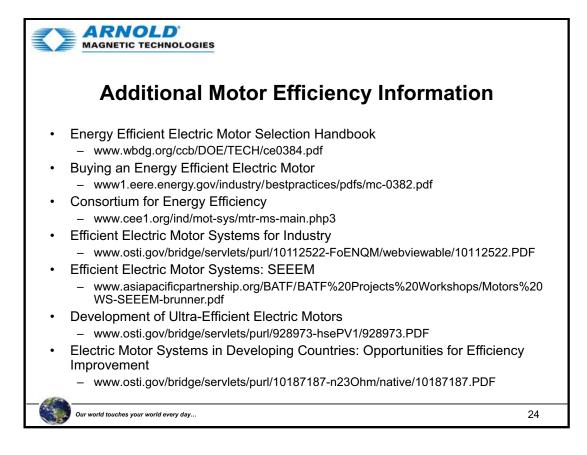
• 20% of energy loss in the motors under this study is from Core Loss, most of which is attributed to the lamination material, either directly or indirectly.



- Despite very large numbers of small motors in use today, large industrial motors consume greater amounts of electric power than smaller motors.
- Large industrial motors are more likely to run continuously, while smaller motors may be on a small fraction of time.

MAGNETIC TECHN							
				• =	[Torrato
Energ	JY P	oncy	Y AC		licie	ncy	Targets
			Nominal Ful			0.00	Increasing Efficiency
Number of Poles	6	OPEN MOTO	2	6	LOSED MOT	2	Targets
Motor Horsepower	-	-				-	
1	80.0	82.5		80.0	82.5	75.5	75.5%
1.5	84.0	84.0	82.5	85.5	84.0	82.5	10.070
2	85.5	84.0	84.0	86.5	84.0	84.0	
3	86.5	86.5	84.0	87.5	87.5	85.5	
5	87.5	87.5	85.5	87.5	87.5	87.5	
7.5	88.5	88.5	87.5	89.5	89.5	88.5	
10	90.2	89.5	88.5	89.5	89.5	89.5	
15	90.2	91.0	89.5	90.2	91.0	902	
20	91.0	91.0	90.2	90.2	91.0	90.2	
25	91.7	91.7	91.0	91.7	92.4	91.0	
30	92.4	92A	91.0	91.7	92.4	91.0	
40	93.0	93.0	91.7	93.0	93.0	91.7	
50	93.0	93.0	92A	93.0	93.0	92.4	
50	93.6	93.6	93.0	93.8	93.6	93.0	
75	93.6	94.1	93.0	93.6	94.1	93.0	
100	94.1	94.1	93.0	94.1	94.5	93.6	95.0%
125	94.1	94.5	93.6	94.1	94.5	94.5	33.078
150	94.5	95.0	93.6	95.0	95.0	94.5	
	94.5	95.0	94.5	95.0	95.0	95.0	

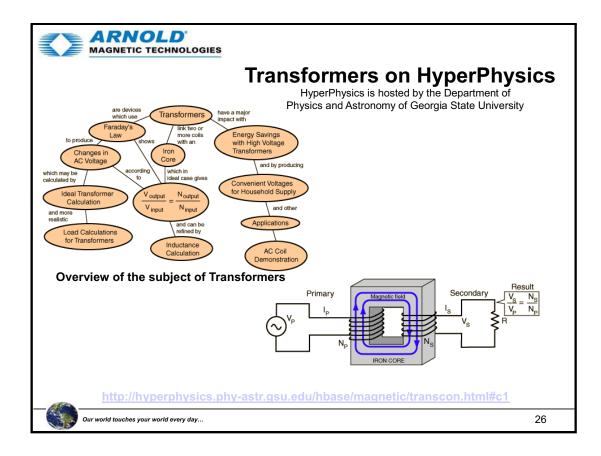
• This chart emphasizes the energy efficiency requirements of higher HP motors since that is where the greatest energy savings is expected to accrue.



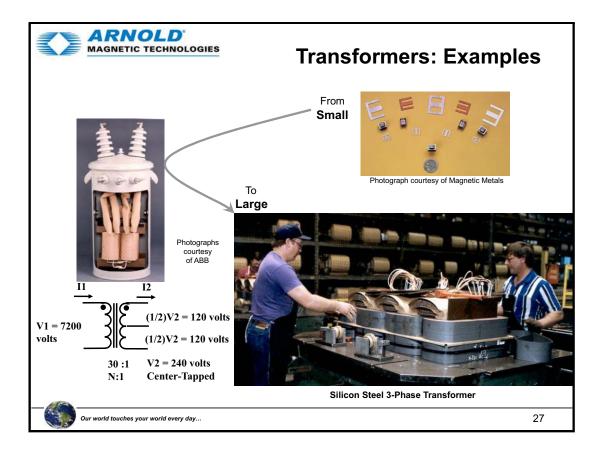
• There is a wealth of information on the internet. Here are just a few additional useful links.



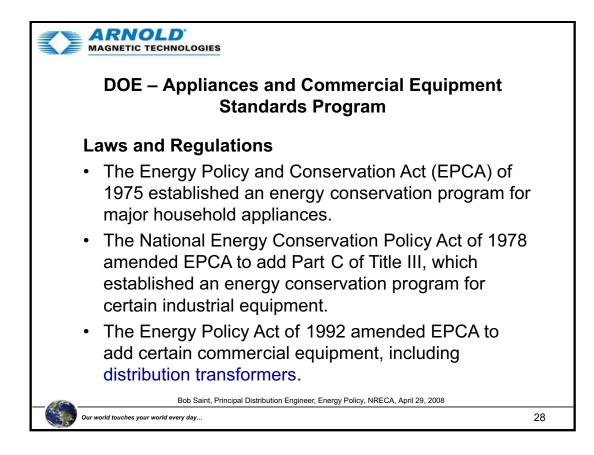
- Transformers and Inductors when used at low frequencies, such as these electric distribution transformers utilize low cost iron laminations and operate with efficiencies of 98% or higher.
- Smaller, special purpose transformers in high frequency and pulse power applications benefit from thin gauge and higher resistivity Si-Fe laminations.
- Examination of this subject would require an entire session by itself, so we will restrict this to just a few slides.

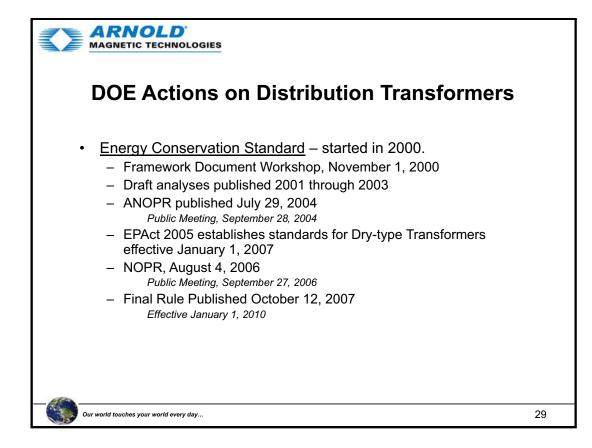


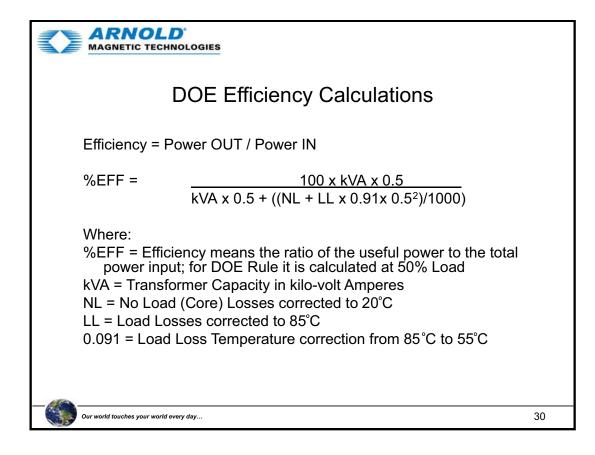
- Transformers are used to convert lower voltages to higher voltages (step-up transformer) or higher voltages to lower ones (step-down).
- The ratio of turns of wire on the primary (input) to those on the secondary (output) determines which type it is.
- Since power stays the same, when voltage increases, current decreases - and the converse is true.
- The applied voltage cannot be steady DC: it must be AC or pulsed or varying DC.
- Each of the windings has associated with it an electrical resistance. Under AC (or pulsed DC) conditions, this resistance become inductive (inductance).
- Therefore, a Si-Fe core with a single winding is an inductor.

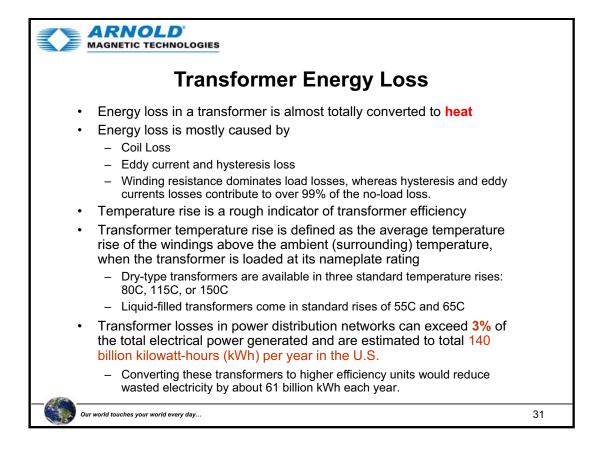


- The picture at the left is a typical step-down transformer which reduces power line voltages to 120/240 volts for domestic use.
- The high voltage connectors are on the top and the 120/240 volt connectors are on the side of the transformer case.









Winding resistance

Current flowing through the windings causes resistive heating of the conductors. At higher frequencies, skin effect and proximity effect create additional winding resistance and losses.

Hysteresis losses

Each time the magnetic field is reversed, a small amount of energy is lost due to hysteresis within the core. For a given core material, the loss is proportional to the frequency, and is a function of the peak flux density to which it is subjected.

Eddy currents

Ferromagnetic materials are also good conductors, and a solid core made from such a material also constitutes a single short-circuited turn throughout its entire length. Eddy currents therefore circulate within the core in a plane normal to the flux, and are responsible for resistive heating of the core material. The eddy current loss is a complex function of the square of supply frequency and inverse square of the material thickness.

Magnetostriction

Magnetic flux in a ferromagnetic material, such as the core, causes it to physically expand and contract slightly with each cycle of the magnetic field, an effect known as magnetostriction. This produces the buzzing sound commonly associated with transformers, and in turn causes losses due to frictional heating in susceptible cores.

Mechanical losses

In addition to magnetostriction, the alternating magnetic field causes fluctuating electromagnetic forces between the primary and secondary windings. These incite vibrations within nearby metalwork, adding to the buzzing noise, and consuming a small amount of power.

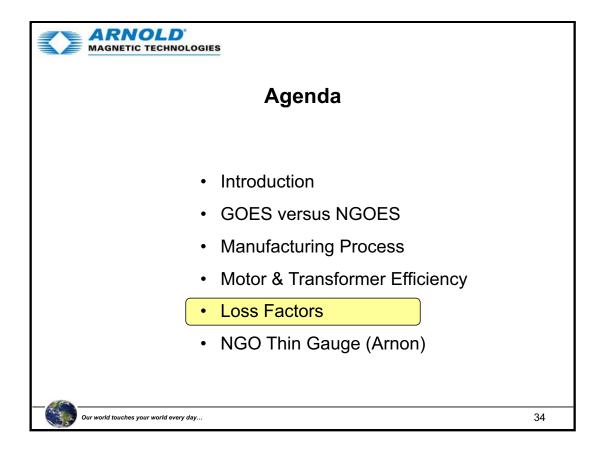
Stray losses

Leakage inductance is by itself lossless, since energy supplied to its magnetic fields is returned to the supply with the next half-cycle. However, any leakage flux that intercepts nearby conductive materials such as the transformer's support structure will give rise to eddy currents and be converted to heat.

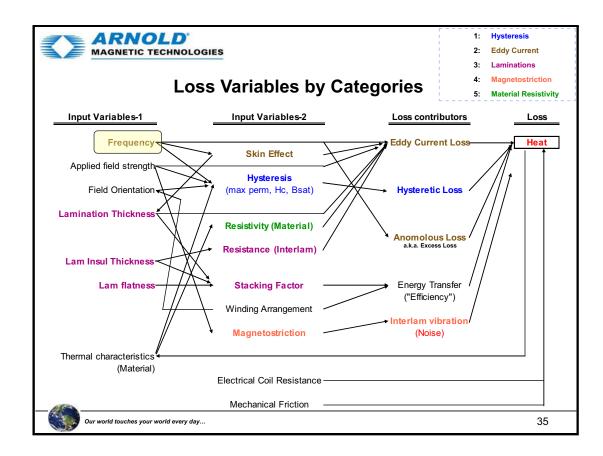
Single Phase Efficiency (%)					
<u>kVA</u>	NEMA TP-1	TSL2	Final Rule	<u>TSL4</u>	
10	98.40	98.40	98.62	98.48	
15	98.60	98.56	98.76	98.63	
25	98.70	98.73	98.91	98.79	
37.5	98.80	98.85	99.01	98.91	
50	98.90	98.90	99.08	99.04	
75	99.00	99.04	99.17	99.08	
100	99.00	99.10	99.23	99.14	
167	99.10	99.21	99.25	99.25	
250	99.20	99.26	99.32	99.45	
333	99.20	99.31	99.36	99.49	
500	99.30	99.38	99.42	99.54	
667	99.40	99.42	99.46	99.57	
833	99.40	99.45	99.49	99.60	

- TSL stands for Trial Standard Levels with #1 through 6.
 - TSL1 = NEMA TP 1-2002 (industry voluntary standard)
 - TSL2 ~1/3 of the efficiency between TP 1 and Min LCC (TSL4)
 - TSL3 ~2/3 of the efficiency between TP 1 and Min LCC (TSL4)
 - TSL4 ~minimum life-cycle cost (LCC)
 - TSL5 ~maximum energy savings with no change in LCC
 - TSL6 = maximum technologically feasible

	Three Phase Efficiency (%)						
<u>kVA</u>	NEMA TP-1	TSL2	Final Rule	TSL4			
15	98.10	98.36	98.36	98.68			
30	98.40	98.62	98.62	98.89			
45	98.60	98.76	98.76	99.00			
75	98.70	98.91	98.91	99.12			
112.5	98.80	99.01	99.01	99.20			
150	98.90	99.08	99.08	99.26			
225	99.00	99.17	99.17	99.33			
300	99.00	99.23	99.23	99.38			
500	99.10	99.32	99.25	99.45			
750	99.20	99.24	99.32	99.37			
1000	99.20	99.29	99.36	99.41			
1500	99.30	99.36	99.42	99.47			
2000	99.40	99.40	99.46	99.51			
2500	99.40	99.44	99.49	99.53			
Our world touches your w	vorld every day			33			



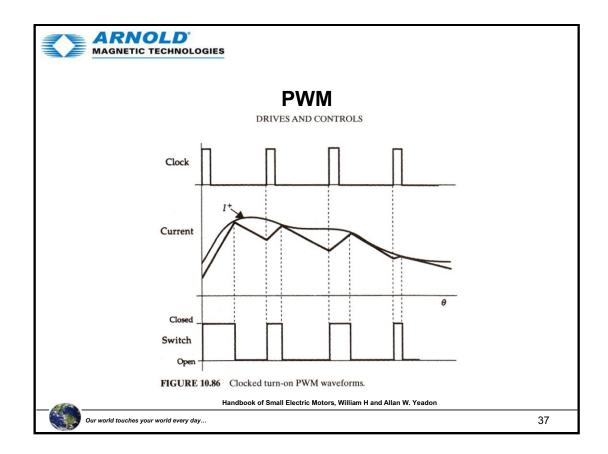
- Si-Fe is more expensive than low carbon steel and thin gauge Si-Fe is more expensive than thick gauge.
- Therefore, to justify the usage of thin gauge Si-Fe, there must be an advantage to counterbalance the higher cost.



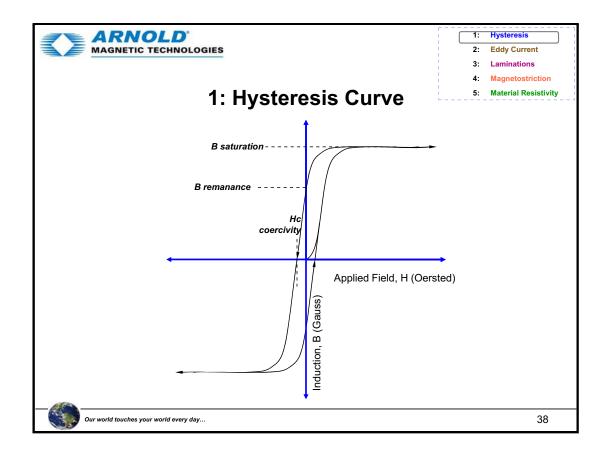
- This slide shows the complex set of variables involved with selecting the proper material grade and thickness.
- Lower efficiency is the result of energy being converted to heat.
- Note that many of the variables are interactive. One variable can affect another variable. For example, switching frequency affects how deep the field will penetrate a lamination which affects required lamination thickness which affects stacking factor, etc.
- That is, the use of thin gauge Si-Fe (Arnon 5 and Arnon 7) can minimize the losses associated with most of these items.
- These variables can be grouped into similar categories for discussion.
- 5 groups have been created here as shown in the upper right of the chart.
- We need to first understand some basics regarding Frequency, so that will be our first topic.

	Table 1.1MaximunCommonly Occurrin	,	rmulas for
	Function	Waveform	Formula
	1. Sine wave voltage (steady-state)	-~	$B_{max} = \frac{E_{rms} \times 10^4}{4.44 NAf}$
Waveforms	2. Symmetrical square wave voltage		$B_{\rm max} = \frac{E_{\rm pk} \times 10^8}{4 N A f}$
	 Interrupted symmetrical square wave voltage 	_ ŢŢr	$B_{\max} = \frac{E_{\rm pk} \times t \times 10^8}{2NA}$
	 Half sine wave voltage pulse 		$B_{max} = \frac{E_{gk} \times 2 \times t \times 10^8}{\pi NA}$
Note: In these formulas N is the number of turns in the winding across which the voltage is developed. A is the	 Unidirectional rectangular voltage pulse 		$B_{\rm max} = \frac{E_{\rm pk} \times t \times 10^8}{NA}$
winding actoss which here voltage is beverbyed. A is the cross-sectional area of the core around which the winding is placed. If the area is expressed in square centimeters, the flux density will be in gauss. If the area is expressed in square inches, the flux density will be in maxwells per square inches, the flux density will be in maxwells per square inch. Time t is in seconds. Frequency <i>f</i> is in hertz. Voltage <i>E</i> in in volts. Current <i>I</i> is in amperes. Inductance <i>L</i> is in henries.	 Full-wave-rectified single-phase sine wave voltage (ac component only) 	A-	$B_{max} = \frac{E_{de} \times 10^8}{19.0 NAf}$
	 Half-wave-rectified three-phase sine wave voltage (ac component only) 	\sim	$B_{\rm max} = \frac{E_{\rm dc} \times 10^8}{75.9 NAf}$
	 Full-wave-rectified three-phase sine wave voltage (ac component only) 	~~~~~	$B_{\rm max} = \frac{E_{\rm dc} \times 10^8}{664 NAf}$
Handbook of Transformer Design & Application, p. 1.6	9. Current	Any	$B_{\rm max} = \frac{LI_{\rm max} \times 10^8}{NA}$
Our world touches your world every day			36

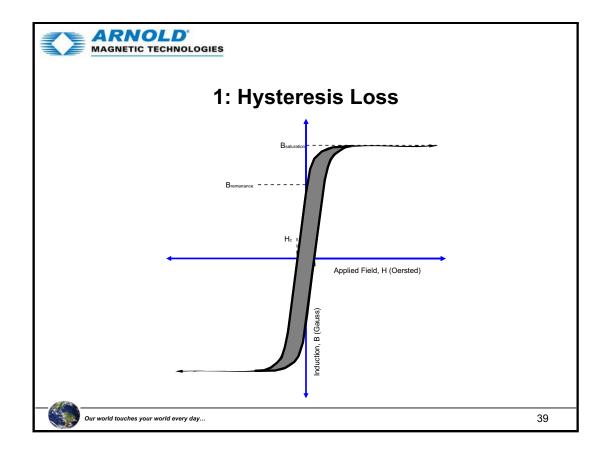
- Before we begin examining magnetic and electrical properties of the material, it is necessary to understand that the affects upon the material are due to the **rate** and **magnitude** of applied field change.
- While we consider a sinusoidal wave form to be most common, developments in electronic controls over the past two decades have allowed application of complex waveforms for both control and efficiency.
- Some of these waveforms produce more rapid field changes than suggested by the over riding frequency. One example of this is a pulse field waveform.



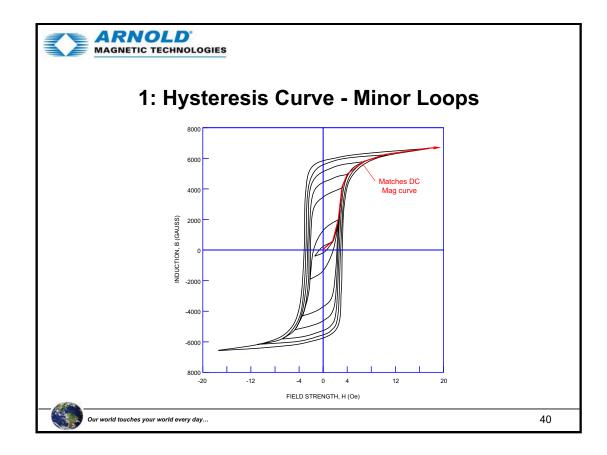
- With Pulse Width Modulation, each pulse exhibits very high frequency (rapid change of electrical field).
- There are many types of PWM including: Clocked Turn on (shown here), Hysteresis, Clock Turn-Off, Dual Current Mode, and Triangle (or sinusoidal) PWM.



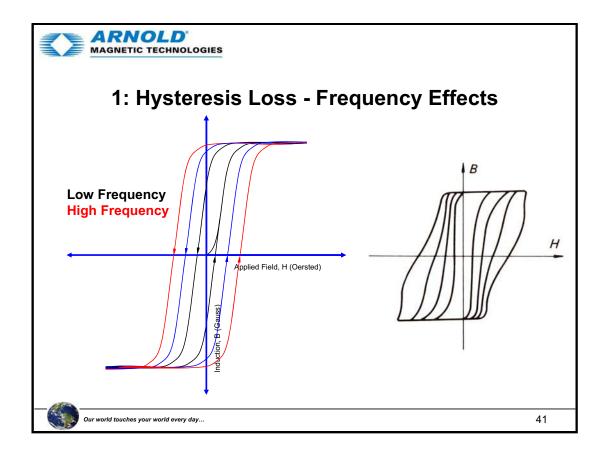
- A sinusoidal magnetic field applied to a ferromagnetic material will drive an applied field (H) back and forth along the horizontal axis as shown in this hysteresis curve. As the applied field increases, the induction increases, etc.
- As it proceeds in the first quadrant, the material can become fully saturated.
- However, as the H field decreases toward the negative region (2nd quadrant), the induction does not follow the exact same path. Instead, there is a lag--or hysteresis in the induction.
- Even when the field reaches zero (H=0), there is still some induction remaining in the material. This is called the remanant flux density, or Br.
- To bring the induction in the material down to zero, the H field must be driven past the origin to a negative value called Hc or HcB -- this is the "coercivity" of the material.
- As the sinusoidal field continues to go through its cycles, this behavior is repeated in both directions (polarities).



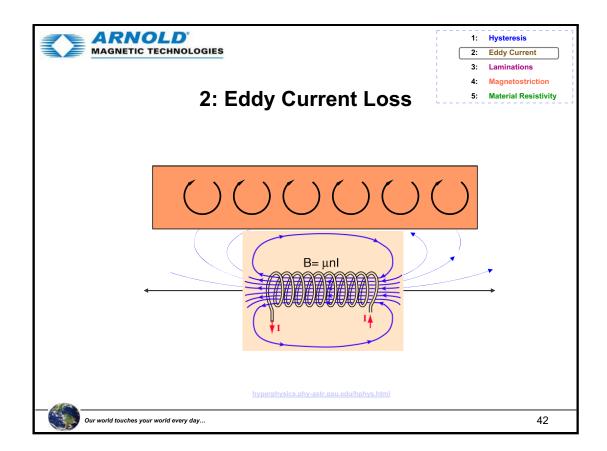
- One implied property from these hysteresis curves is that there is significant energy absorption, represented by the area inside the hysteresis curve.
- In soft magnetic applications, this energy absorption manifests itself as "hysteresis loss".
- The hysteresis loss is the energy absorbed by the material as it being magnetized in one direction, magnetized in the reverse direction, and re-magnetized in the original direction.



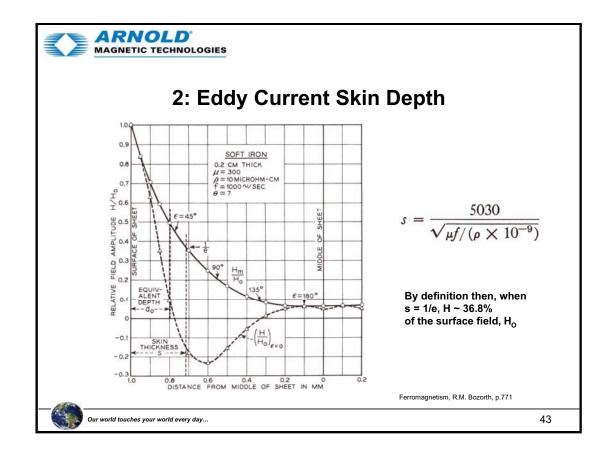
- In a PM motor, the material is not likely to be driven to saturation except at the pole tips. As a result, much of the material will be using "minor" loop properties.
- This is a series of super-imposed minor hysteresis loops. It can be clearly seen that the induction response differs greatly depending upon the level of applied field (H).
- The energy loss will be equal to the area within the specific minor loop experienced by each part of the lamination.
- Since not all the material will see the same externally applied field, there will be unequal energy loss from place to place within the lamination structure.



- Additionally, when the frequency of the applied field is increased, it has been found that hysteresis loss also increases.
- This is manifested as an increase in the area within the hysteresis curve(s). (See the left chart).
- As the frequency continues to increase, however, another aspect of core loss begins to appear, causing not only an enlargement of the area inside the curve, but also distortion.
- This distortion is due in large part to eddy currents.

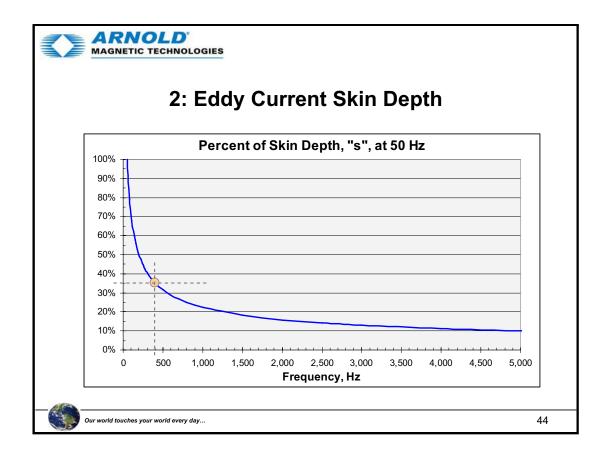


- As the magnetic field alternates through a conductor, eddy currents are created in the material. These are localized currents, flowing in a closed path within the lamination. In accordance with Lenz's law, these currents oppose change in the field that is inducing the eddy currents.
- The creation of these currents requires energy, and therefore is a source of energy loss.
- Eddy current loss is in addition to hysteresis loss.
- It should be noted that eddy currents can occur in any conductive material including non-magnetic copper and aluminum. In fact, it can be shown that eddy currents in a non-magnetic conductor can exert significant braking effect relative to a moving magnet (forming the basis for hysteresis brakes, clutches and drives).
- The key point here is that eddy currents are caused either by a permanent magnet moving in relation to the conductive material or by a changing electromagnetic field generated by current flow in a conductor.

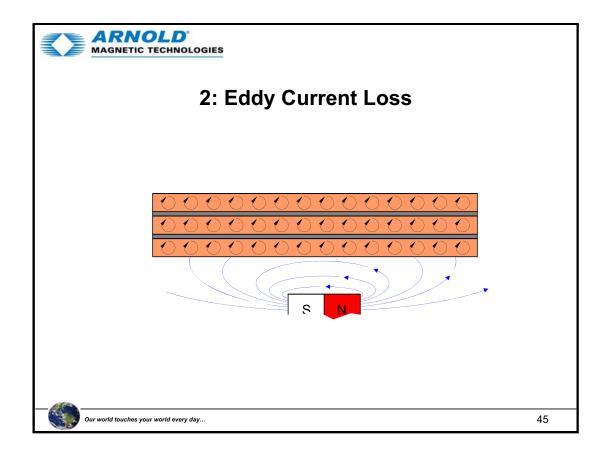


- As frequency increases, eddy currents cannot penetrate as far into the conductive material. This is known as skin effect.
- · Bozorth shows the skin depth, "s", as a function of

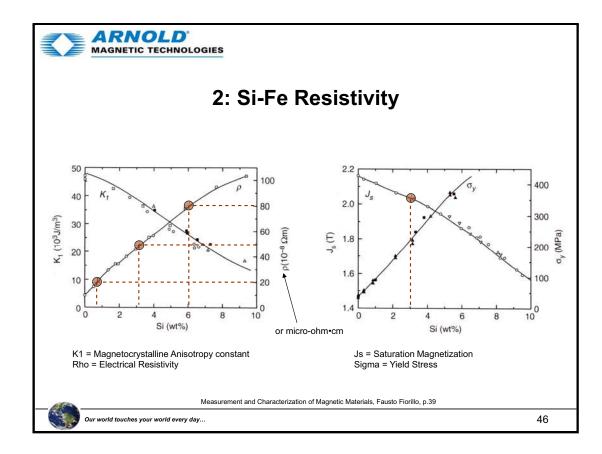
f = frequency in Hz Mu = permeability Rho = resistivity in ohm•cm



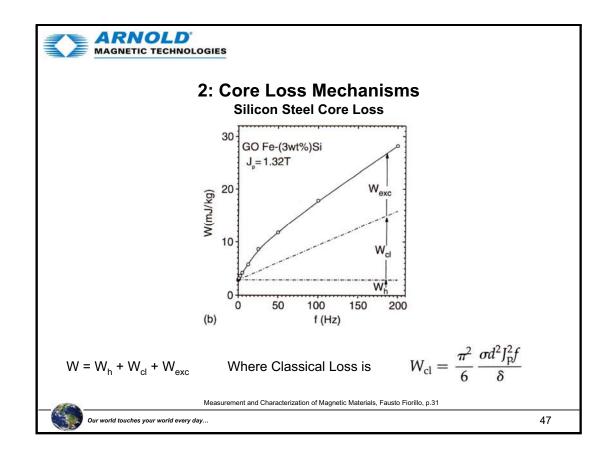
- As frequency increases eddy currents penetrate to a far lesser extent.
- Penetration depth decreases very rapidly when (switching) frequency increases above 50 Hz, reaching at 400 Hz just ~35% of the penetration depth at 50 Hz .



- Since eddy currents are caused by induced voltage in conductive material, dividing the material into insulated layers breaks up the induced voltage lowering eddy current loss dramatically. That is: the sum of the losses per layer is much less than the loss for the undivided material. Induced voltage is proportional to frequency so there is greater benefit from more and thinner layers as the frequency increases.
- In practical devices then, eddy current loss can be lowered by reducing the thickness of the laminations and including an insulating coating between the laminations.
- In fact, coating technology plays a major role in the ability to use soft magnetic materials at the high frequencies becoming common today.
- A second method to achieve reduction in eddy current loss is use of materials with relatively high resistivity.
- Incidentally, this is why ferrite cores, being a ceramic insulator, are used in very high frequency transformer/inductor applications. However, their saturation magnetization is so low that they are not considered useful for motors and generators.
- Another family of materials, SMC's (Soft Magnetic Composites) are of interest for frequencies above those where thin gauge Si-Fe is useful.

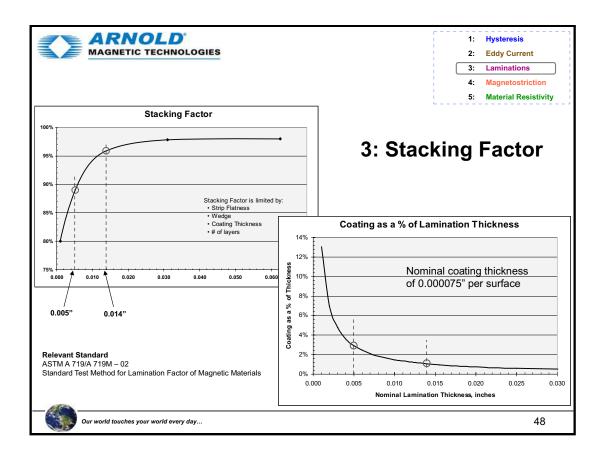


- The electrical resistivity of 3% Silicon-iron is approximately 48 micro-ohm• cm.
- This is about five times higher than low carbon steel.
- Saturation magnetization is reduced by the presence of silicon, dropping by about 6% with 3% Si. But since most of the lamination is not driven to saturation in PM motors, this small drop has minimal affect on normal use.

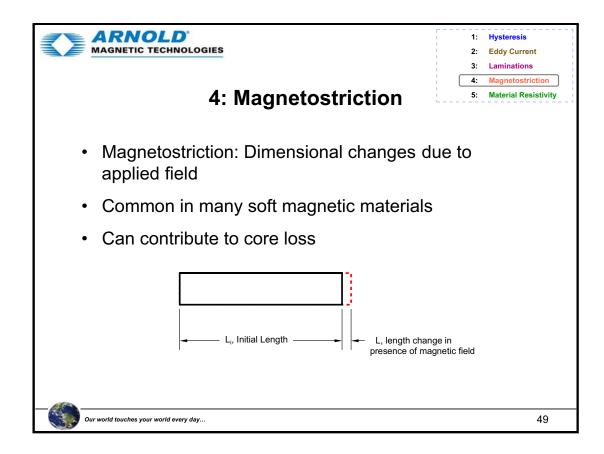


- This chart shows the actual measured core loss of a silicon iron lamination structure with loss components.
- Note that in reference literature, the Hysteresis Loss (Wh) is close to constant with frequency. The distortion of the hysteresis loop and the resulting shifts in loss may be ascribed to the eddy currents.
- The Classical Loss (Wcl, including eddy current loss) is shown as proportional to frequency.
- The remaining loss is complex in mechanism, and is often referred to as: Anomalous Loss, Residual Loss or Excess Loss. It has not yet been satisfactorily explained though efforts are underway to do so such as the efforts at Clarkson University under the direction of Professor Prag Pillay.
- Lamination thickness in this example is 0.29 mm under sinusoidal time dependence of polarization.

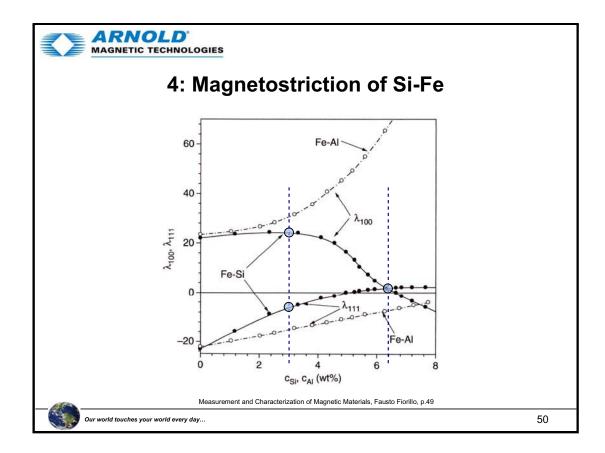
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Delta = density
d = lamination thickness
Sigma = conductivity
Jp = Peak Polarization
f = frequency
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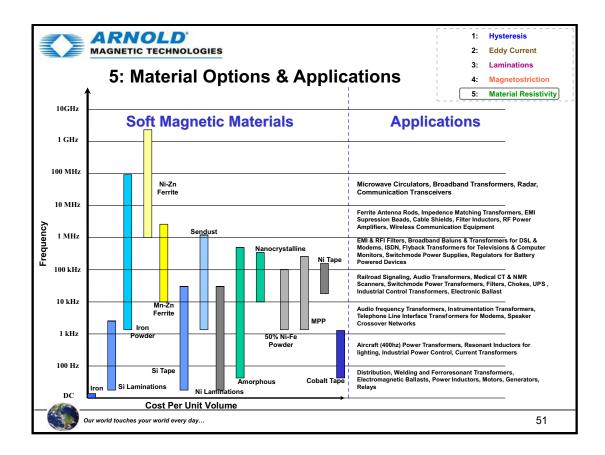
- We've seen that reducing lamination thickness can reduce core loss due to eddy currents. However, there are several drawbacks that provide practical limits to how thin the laminations may be economically made and used.
 - 1. **COATING**: First, the coating, although very thin (between 50 and 120 millionths of an inch), separates the laminations. We see in the chart to the right that with 0.005" laminations that a typical coating represents 3% of the total lamination thickness. Were there no other effects, the stacking factor could be a maximum of only 97%.
 - 2. **PHYSICAL IMPERFECTIONS**: Secondly, laminations are not physically perfect. They suffer from wedge and waffling, both contribute to irregular gaps between layers and the actual achievable stacking factor is approximately as shown in the left chart.
 - 3. **QUANTITY HANDLED**: Another issue is that of requirements for punching and stacking larger numbers of laminations to make up the total stack height.
 - 4. **PUNCH DIFFICULTY**: A 4th issue is that punching of steel becomes more difficult as the laminations become thinner. Thicknesses to 0.001" are possible, but punch difficulty increases dramatically below a lamination thickness of 0.005". At these thin gauges punching is actually a controlled tear. Punch tooling has to be of the highest quality and maintained rigorously.
- The use of thin gauge, therefore, depends upon a balance of performance with cost and technical feasibility. Switching frequencies of 400 Hz and higher or the requirement for extremely low core loss make thin gauge NGOES a requisite.



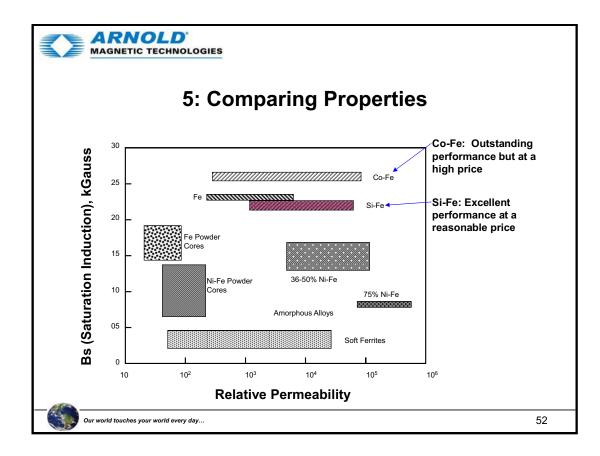
- Ferromagnetic materials experience minor dimensional change when placed in a magnetic field. This effect is called "magnetostriction."
- The magnitude of these physical changes (L/Li) are small, typically a few parts per million.
- These dimensional changes absorb energy and contribute to the total core loss.
- Magnetostriction can manifest itself as audible noise, e.g. "transformer hum."



- Magnetostriction of 3% Si-Fe varies from about 23 in the easy axis to a negative 6 in the hard axis of orientation - both numbers x10E-6.
- At ~6.5% Silicon, magnetostriction reaches a minimum. However, at this level of silicon, the alloy is too hard to roll successfully, especially to thin gauge.
- (I have not yet found data for magnetostriction of NGOES).



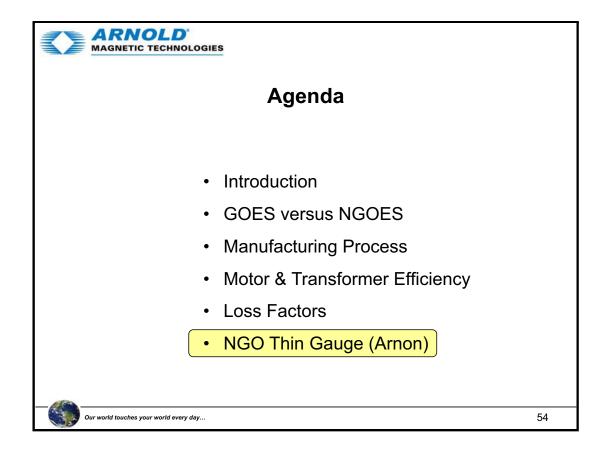
- This chart offers a comparison of soft magnetic materials as a function of cost and usable frequency range.
- Maximum use frequency depends on the operating flux density so the chart should be viewed as a general guide.
- Increasing cost per unit volume is shown from left to right. Performance improvements from a more expensive core material can provide for lower overall system cost.
- Silicon steel is used so widely because itrepresents a good comb ination of cost and performance.



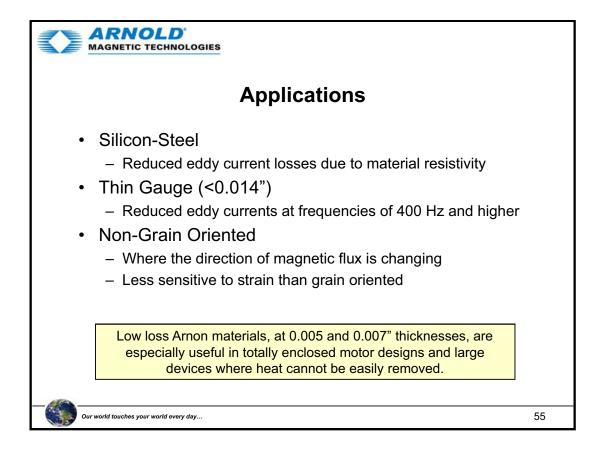
- This slide plots permeability versus saturation induction for common soft magnetic materials. Both properties are usually desirable, so the best performing material shown would be the iron-cobalt alloys. While they are expensive, they are among of the best performers.
- Only slightly lower in performance is the much less costly Si-Fe material.
- Although iron can have very high permeability and high magnetic saturation, it must be pure, annealed properly and is still susceptible to hardening from mechanical working and aging.

	IEC	EN	1	ASTM	JIS	GOST
	404-8-4	10106	AISI	A677	2552	21427
	-1986	-1995		-1989	-1986	0-75
5: Silicon Steel	-	M235-50A	-	-	-	-
– • •	250-35-A5	M250-35A	M 15	36F145	35A250	2413
Designations	270-35-A5	M270-35A	M 19	36F158	35A270	2412
	300-35-A5	M300-35A	M 22	36F168	35A300	2411
	330-35-A5	M330-35A	M 36	36F190	-	-
The American Iron and Steel Institute type	-	M250-50A	-	-	-	-
numbers for electrical steel grades consist	270-50-A5	M270-50A	-	-	50A270	-
of the letter M followed by a number. The	290-50-A5	M290-50A	M 15	47F168	50A290	2413
A stands for magnetic material; the	310-50-A5	M310-50A	M 19	47F174	50A310	2412
number is representative of the core loss	330-50-A5	M330-50A	M 27	47F190	-	-
or that grade. At the time the AISI system	350-50-A5	M350-50A	M 36	47F205	50A350	2411
vas adopted, the type number assigned to	400-50-A5	M400-50A	M 43	47F230	50A400	2312
each grade was approximately 10 times	470-50-A5	M470-50A	-	47F280	50A470	2311
he core loss for 29 gauge material at 15	530-50-A5	M530-50A	M 45	47F305	-	2212
G and 60 Hz.	600-50-A5	M600-50A	-	-	50A600	2112
	700-50-A5	M700-50A	M 47	47F400	50A700	-
oday, the type numbers do not have the	800-50-A5	M800-50A	-	47F450	50A800	2111
specific association with core loss	-	M940-50A	-	-	-	-
	-	M310-65A	-	-	-	-
ecause electrical steels have been	-	M330-65A	-	-	-	-
significantly improved, and the core loss	350-65-A5	M350-65A	M 19	64F208	-	-
juarantees substantially reduced.	400-65-A5	M400-65A	M 27	64F225	-	-
However, the numbers do indicate not only	470-65-A5	M470-65A	M 43	64F270	-	-
a specific grade but also the relative core	530-65-A5	M530-65A	-	-	-	2312
osses of grades within a class.	600-65-A5	M600-65A	M 45	64F360	-	2212
	700-65-A5	M700-65A	-	64F400	-	2211
AK Steel promotional literature: 24-784	800-65-A5	M800-65A	-	-	65A800	2112
	- 1000-65-A5	- M1000-65A	M 47	64F500 64F550	- 65A1000	

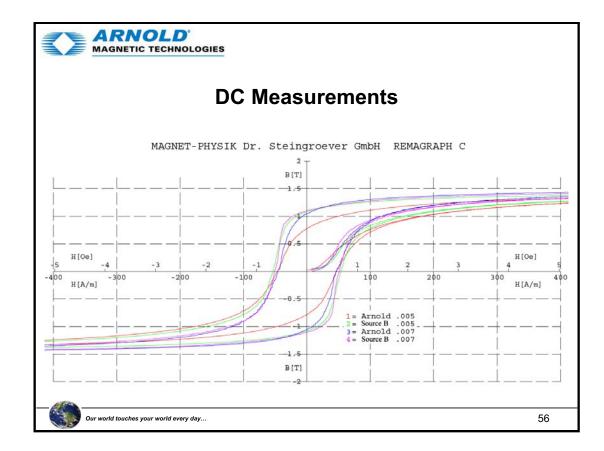
• There is not "1" silicon iron material, but a whole family of materials: oriented and not oriented, different thicknesses and coatings, varying silicon contents, different processing, etc.



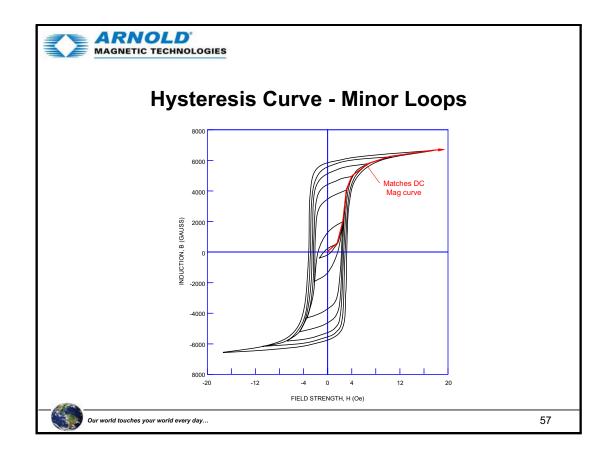
- Thin gauge Si-Fe (<=0.007", e.g. Arnon) is particularly useful for high frequency operation to reduce eddy current loss.
- Oriented product is commonly used in high frequency or burst pulse transformer applications where the benefit of directional properties can be utilized.
- In motors, generators and rotating machinery in general, non-oriented Si-Fe is preferred since the field intersects the material at varying angles and performance benefits from material isotropy.



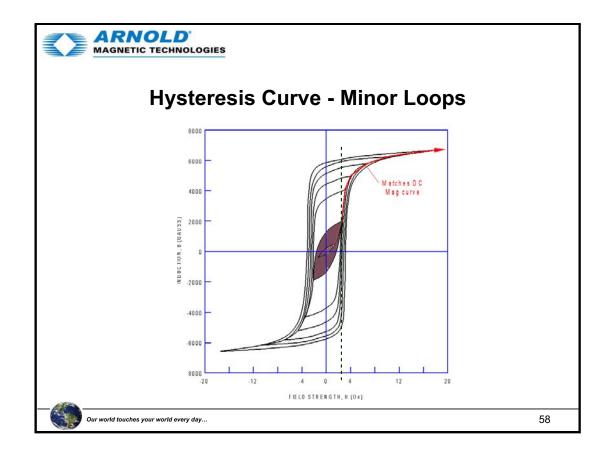
- Three good reasons to use Arnon are:
 - Silicon steel has higher inherent resistivity
 - Thin gauge reduces eddy current loss
 - Non-oriented magnetic properties improve performance in rotating machinery
- Arnon HcB is less than 0.6 Oersteds, 8 to 15% less than competitive NGOES, resulting in lower hysteresis core loss. (Comparative measurements on same equipment during the same measurement session). Total loss is reported to be as much as 50% lower than for competitive thin gauge materials.
- This lower HcB results from very specific processing conditions during rolling and annealing.



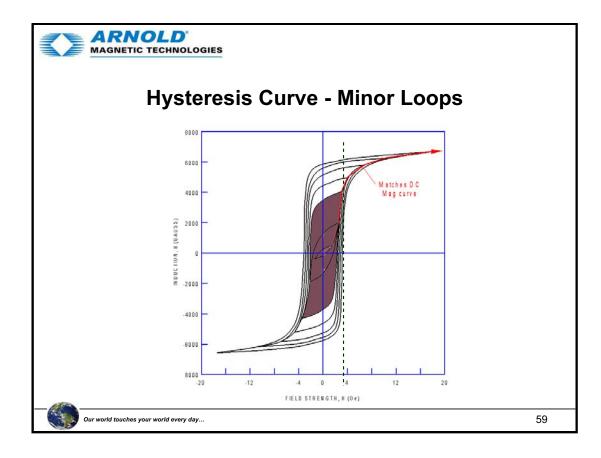
- Recent measurements or Arnon 5 and 7 mil product versus other commercial products is shown in these DC hysteresis loops.
- In addition to the different loop shapes, Arnon has consistently lower values for HcB (18% lower for Arnon 5 versus competitive materials).

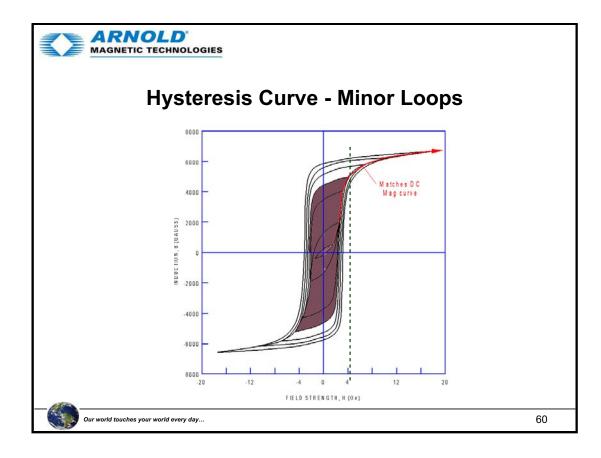


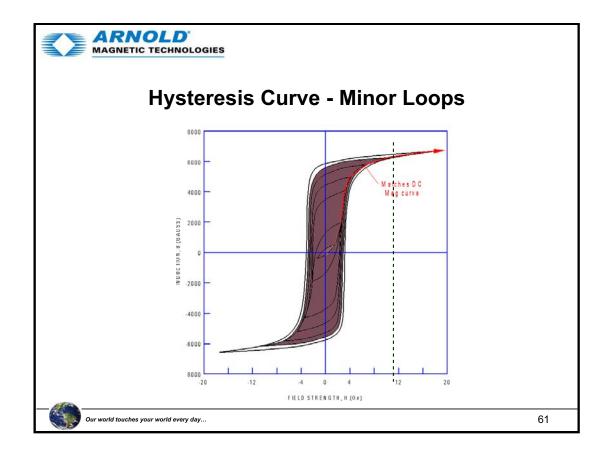
- In a PM motor, the material is unlikely to be driven to saturation except at the pole tips. As a result, most of the material will be using "minor" loop properties.
- This is a series of super-imposed minor hysteresis curves. It can be clearly seen that the induction response differs greatly depending upon the level of applied field (H).
- The energy loss will be proportional to the area within the specific minor loop within the lamination material.
- Since not all the material will see the same externally applied field, there will be unequal energy loss from place to place within the lamination structure.
- The more square-loop the material, the greater the sensitivity to small changes in applied field.



- The more square loop the material, the greater the sensitivity to small changes in applied field as we see in this series of illustrations.
- The dashed green line marks the applied field, H.



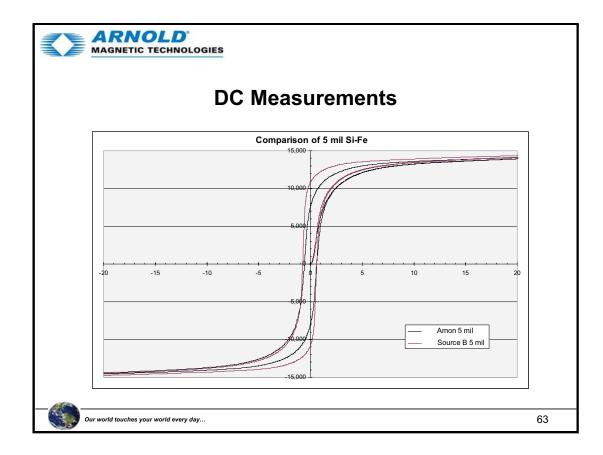




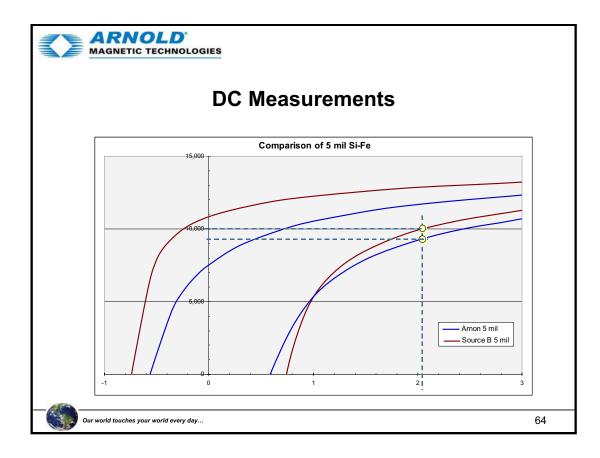
- It is apparent that initially large changes in loss occur.
- Then, as the material nears saturation, large changes in applied field are required to make noticeable changes in loss.

		ETIC TI		LOGIES										
								5 mil Si-F		;]
							^{20,000} F	• •	•					
							15,000							
							10,000							
							-5,000							
	-600	-500	-400	-300	-200	-100	• • • • •	100	200	300	400	500	600	
							-5,000						_	
							10,000				Arnon 5	mil		
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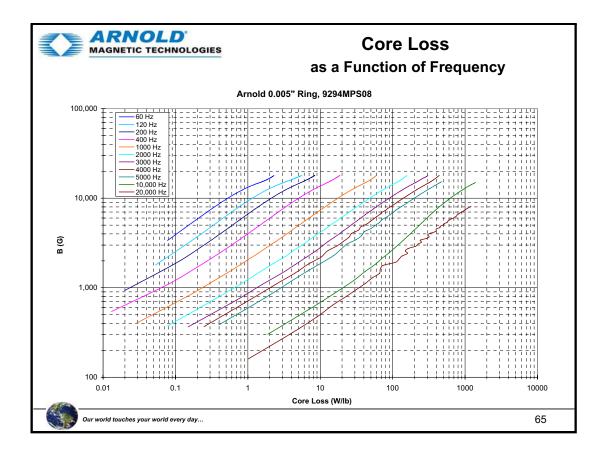
• Actual DC hysteresis curves for the materials are shown in the next series of charts going from full to magnified view.



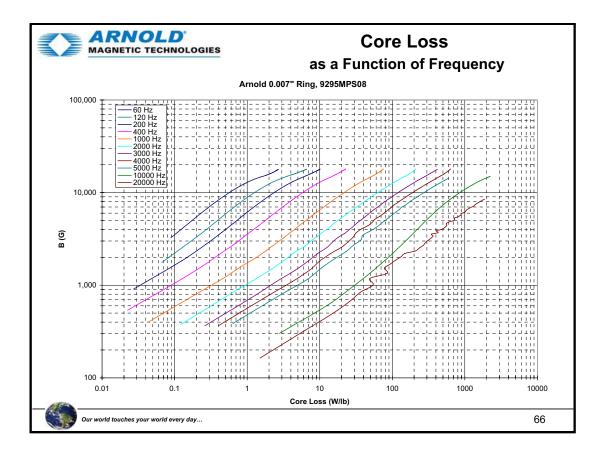
• In this second chart the difference between Arnon and the competitive material is clearly evident.



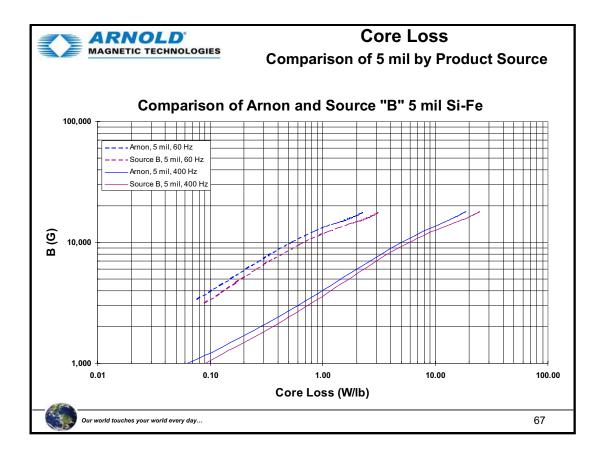
- What we see is the difference in "B" achieved at any level of "H".
- Specifically, The applied field to drive the competitor material to B=10,000 only drives the Arnon to 9200 Gauss.
- At the pole tips both materials will be driven to saturation. However, at locations of lower applied field, the Arnon will not be driven to as high a "B."
- While we could integrate within the curves to calculate the difference in hysteresis loss, it is easier to simply measure it.



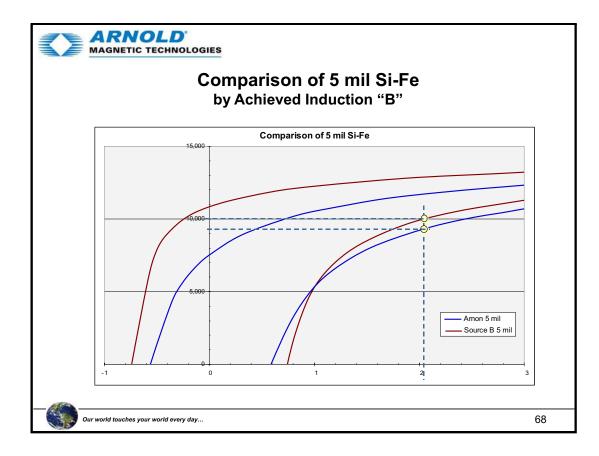
- Core Loss at various frequencies can be measured.
- This chart is for Arnon 5 and for frequencies between 60 and 20,000 Hz.



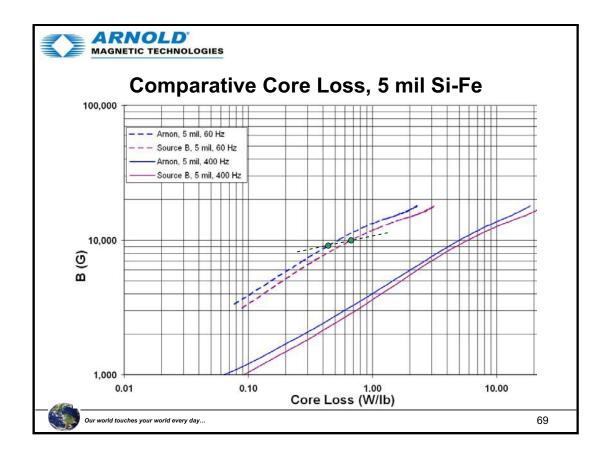
• This presents the same data for Arnon 7.



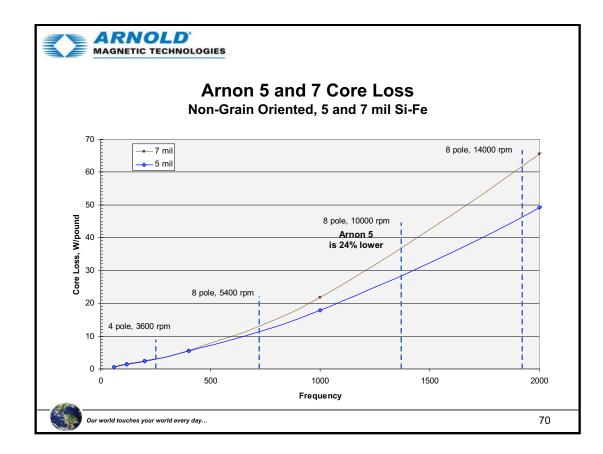
• A comparison between Arnon 5 and the source B 5 mil material is shown here.



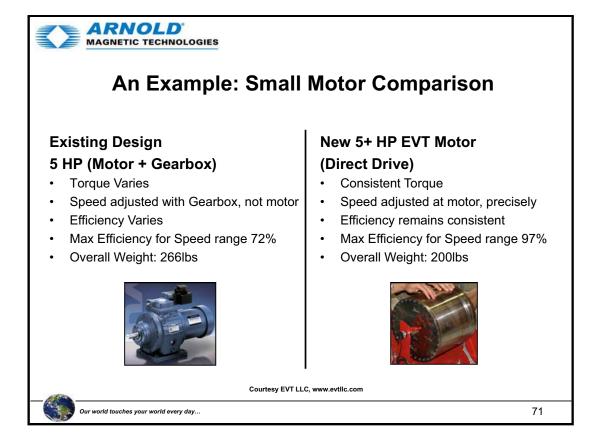
• By referring to this chart we see that when Source B material is driven to 10,000 Gauss, that the Arnon is driven to 9300.



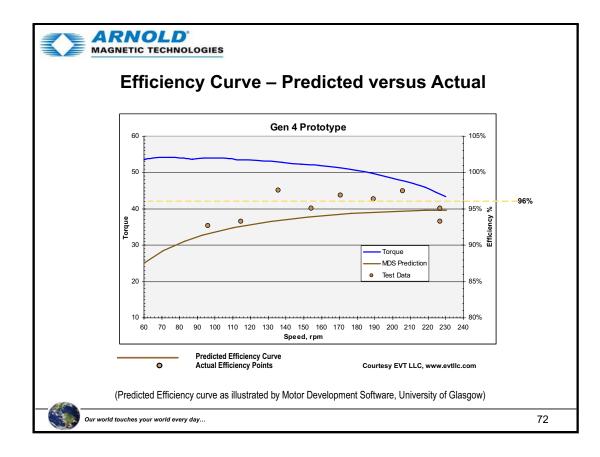
- By comparing the core loss at the level of "B" achieved by the applied field in the device, we see a core loss difference of 0.45 (Arnon) versus 0.68 (Source B).
- Source B is 50 % higher than the Arnon when driven by the same "H" field.
- This confirms anecdotal customer information on the relative performance.
- It confirms the importance of hysteresis curve shape as well as saturation magnetization.



- Below 400 Hz, core loss for both Arnon 5 and Arnon 7 are similarly low.
- Above 400 Hz, Arnon becomes increasingly more advantageous.
- Several example frequencies are shown here emphasizing that higher RPM and greater number of poles drives the switching frequency higher.



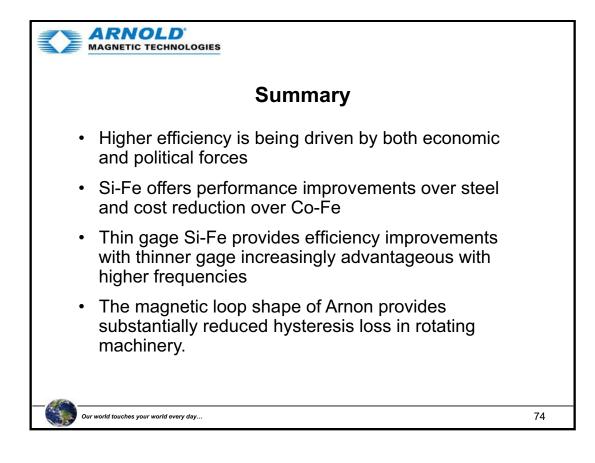
• In this example, the challenge is to replace an open type motor and gear box with a **totally sealed** motor without gear box while improving on the linearity of torque and raising the efficiency.

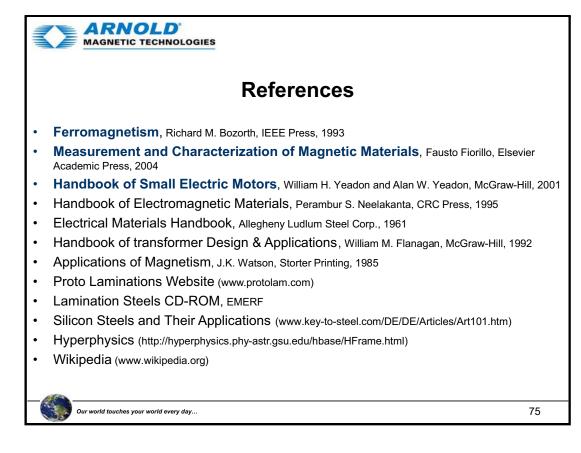


- Data is from 80 to 230 output rpm.
- The torque curve is exceptionally flat, especially at the lower rpm's.
- Efficiency averages 96% over more than half the speed range and peaks between 97 and 98%
- This drive, which uses Arnon 7, is totally enclosed with no cooling and experiences ~10 °C rise above ambient in continuous operation.

ARNOLD MAGNETIC TECHNOLOGIES									
١	/ery Thir	n Gauge Si-	Fe (Grai	n-Orient	ed)				
GC	DES (Grain orie	NGOES is used for nted electrical steel)	0		luctors.				
Table 1. Recommended Grain Oriented Silicon Steel Table 2. Recommended Grain Oriented Silicon Steel Thicknesses for Various Operating Frequencies Table 2. Recommended Grain Oriented Silicon Steel									
	Recommended	Approximate Induction for 300 mW/cc	Pulse Width	Recommended Thickness	Pulses per Second				
Frequency 400 Hz 1 kHz	Thickness 4 mil or 6 mil 4 mil or 6 mil	18 W/lb, 40 W/kg* 15,000 G* 10,000 G	2 to 1000 microseconds	4 mil or 6 mil (D-U, U-I, L-L Laminations)	to 1000				
2 kHz 5 kHz	2 mil 1 mil	6,000 G 3.000 G	0.25 to 2 microseconds	1 mil or 2 mil (C-core)	to 1000				
	C-cores). At 400 Hz, ma	data records. (Arnold no gnetizing current limits the			data records. (Arnold no gnetizing current limits the				
A States									

- Arnold also produces **grain oriented** Si-Fe at 1, 2, 4, and 6 mil, primarily for use in laminated and wound high frequency transformers and chokes.
- Although Grain-Oriented is not normally used for rotating machinery, at least one recent design incorporates this material in a segmented stator design - segmented to allow alignment of the laminations more closely to the applied field.





• All of these were useful, but the three in dark blue were excellent.

